7. ELECTROMAGNETIC RELATIONS

Revised September 2005 by H.G. Spieler (LBNL).

Quantity	Gaussian CGS	SI
Conversion factors:		
Charge:	$2.997\ 924\ 58 \times 10^9 \ esu$	= 1 C = 1 A s
Potential:	(1/299.792458) statvolt (ergs/esu)	= 1 V = 1 H S = 1 V = 1 J C ⁻¹
Magnetic field:	10^4 gauss = 10^4 dyne/esu	$= 1 \text{ T} = 1 \text{ N} \text{ A}^{-1} \text{m}^{-1}$
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	$\mathbf{F} = q \left(\mathbf{E} + \frac{\mathbf{v}}{c} \times \mathbf{B} \right)$	$\mathbf{F} = q\left(\mathbf{E} + \mathbf{v} \times \mathbf{B}\right)$
	$\nabla \cdot \mathbf{D} = 4\pi\rho$	$\nabla \cdot \mathbf{D} = \rho$
	$\mathbf{ abla} imes \mathbf{H} - rac{1}{c} rac{\partial \mathbf{D}}{\partial t} = rac{4\pi}{c} \mathbf{J}$	$\mathbf{\nabla} imes \mathbf{H} - rac{\partial \mathbf{D}}{\partial t} = \mathbf{J}$
	$\nabla \cdot \mathbf{B} = 0$	$\nabla \cdot \mathbf{B} = 0$
	$\boldsymbol{\nabla} \times \mathbf{E} + \frac{1}{c} \frac{\partial \mathbf{B}}{\partial t} = 0$	$\boldsymbol{\nabla} \times \mathbf{E} + \frac{\partial \mathbf{B}}{\partial t} = 0$
Constitutive relations:	$\mathbf{D} = \mathbf{E} + 4\pi \mathbf{P}, \mathbf{H} = \mathbf{B} - 4\pi \mathbf{M}$	$\mathbf{D} = \epsilon_0 \mathbf{E} + \mathbf{P}, \mathbf{H} = \mathbf{B}/\mu_0 - \mathbf{M}$
Linear media:	$\mathbf{D} = \epsilon \mathbf{E}, \mathbf{H} = \mathbf{B}/\mu$	$\mathbf{D} = \epsilon \mathbf{E}, \mathbf{H} = \mathbf{B}/\mu$
	1	$\epsilon_0 = 8.854 \ 187 \ldots \times 10^{-12} \ \mathrm{F} \ \mathrm{m}^{-1}$
	1	$\mu_0 = 4\pi \times 10^{-7} \text{ N A}^{-2}$
	$\mathbf{E} = -\boldsymbol{\nabla}V - \frac{1}{c}\frac{\partial \mathbf{A}}{\partial t}$	$\mathbf{E} = -\boldsymbol{\nabla}V - \frac{\partial \mathbf{A}}{\partial t}$
	$\mathbf{B} = \mathbf{ abla} imes \mathbf{A}$	$\mathbf{B} = \mathbf{ abla} imes \mathbf{A}$
	$V = \sum_{\text{charges}} \frac{q_i}{r_i} = \int \frac{\rho\left(\mathbf{r}'\right)}{\left \mathbf{r} - \mathbf{r}'\right } \ d^3x'$	$V = \frac{1}{4\pi\epsilon_0} \sum_{\text{charges}} \frac{q_i}{r_i} = \frac{1}{4\pi\epsilon_0} \int \frac{\rho\left(\mathbf{r}'\right)}{\left \mathbf{r} - \mathbf{r}'\right } d^3x'$
	$\mathbf{A} = \frac{1}{c} \oint \frac{I \mathbf{d}\boldsymbol{\ell}}{ \mathbf{r} - \mathbf{r}' } = \frac{1}{c} \int \frac{\mathbf{J}(\mathbf{r}')}{ \mathbf{r} - \mathbf{r}' } d^3x'$	$\mathbf{A} = \frac{\mu_0}{4\pi} \oint \frac{I \mathbf{d}\boldsymbol{\ell}}{ \mathbf{r} - \mathbf{r}' } = \frac{\mu_0}{4\pi} \int \frac{\mathbf{J}(\mathbf{r}')}{ \mathbf{r} - \mathbf{r}' } d^3 x'$
	$\mathbf{E}_{\parallel}'=\mathbf{E}_{\parallel}$	$\mathbf{E}_{\parallel}'=\mathbf{E}_{\parallel}$
	$\mathbf{E}_{\perp}' = \gamma(\mathbf{E}_{\perp} + \frac{1}{c}\mathbf{v}\times\mathbf{B})$	$\mathbf{E}_{\perp}' = \gamma(\mathbf{E}_{\perp} + \mathbf{v} \times \mathbf{B})$
	$\mathbf{B}_{\parallel}'=\mathbf{B}_{\parallel}$	$\mathbf{B}_{\parallel}'=\mathbf{B}_{\parallel}$
	$\mathbf{B}_{\perp}' = \gamma (\mathbf{B}_{\perp} - \frac{1}{c} \mathbf{v} \times \mathbf{E})$	$\mathbf{B}_{\perp}' = \gamma(\mathbf{B}_{\perp} - \frac{1}{c^2}\mathbf{v} \times \mathbf{E})$
$\frac{1}{4\pi\epsilon_0} = c^2 \times 10^{-7} \text{ N A}^{-2} = 8.98755 \times 10^9 \text{ m F}^{-1} \text{ ; } \frac{\mu_0}{4\pi} = 10^{-7} \text{ N A}^{-2} \text{ ; } c = \frac{1}{\sqrt{\mu_0\epsilon_0}} = 2.99792458 \times 10^8 \text{ m s}^{-1}$		

7.1. Impedances (SI units)

 ρ = resistivity at room temperature in 10⁻⁸ Ω m:

$$\sim 1.7 \text{ for Cu} \sim 5.5 \text{ for W}$$

$$\sim 2.4 \text{ for Au} \sim 73 \text{ for SS } 304$$

$$\sim 2.8 \text{ for Al} \sim 100 \text{ for Nichrome}$$

(Al alloys may have double the Al value.)

For alternating currents, instantaneous current I, voltage V, angular frequency $\omega :$

$$V = V_0 e^{j\omega t} = ZI . ag{7.1}$$

Impedance of self-inductance $L: Z = j\omega L$.

Impedance of capacitance $C: Z = 1/j\omega C$.

Impedance of free space: $Z = \sqrt{\mu_0/\epsilon_0} = 376.7 \ \Omega$.

High-frequency surface impedance of a good conductor:

$$Z = \frac{(1+j) \rho}{\delta}, \quad \text{where } \delta = \text{skin depth} ; \qquad (7.2)$$

$$\delta = \sqrt{\frac{\rho}{\pi\nu\mu}} \approx \frac{6.6 \text{ cm}}{\sqrt{\nu \text{ (Hz)}}} \quad \text{for Cu} . \tag{7.3}$$

7.2. Capacitors, inductors, and transmission Lines

The capacitance between two parallel plates of area A spaced by the distance d and enclosing a medium with the dielectric constant ε is

$$C = K \varepsilon A/d$$
, (7.4)

where the correction factor K depends on the extent of the fringing field. If the dielectric fills the capacitor volume without extending beyond the electrodes. the correction factor $K \approx 0.8$ for capacitors of typical geometry.

The inductance at high frequencies of a straight wire whose length ℓ is much greater than the wire diameter d is

$$L \approx 2.0 \left[\frac{\mathrm{nH}}{\mathrm{cm}} \right] \cdot \ell \left(\ln \left(\frac{4\ell}{d} \right) - 1 \right) .$$
 (7.5)

For very short wires, representative of vias in a printed circuit board, the inductance is

$$L(\text{in nH}) \approx \ell/d$$
. (7.6)

A transmission line is a pair of conductors with inductance L and capacitance C. The characteristic impedance $Z = \sqrt{L/C}$ and the phase velocity $v_p = 1/\sqrt{LC} = 1/\sqrt{\mu\varepsilon}$, which decreases with the inverse square root of the dielectric constant of the medium. Typical coaxial and ribbon cables have a propagation delay of about 5 ns/cm. The impedance of a coaxial cable with outer diameter D and inner diameter d is

$$Z = 60 \ \Omega \cdot \frac{1}{\sqrt{\varepsilon_r}} \ln \frac{D}{d}, \tag{7.7}$$

where the relative dielectric constant $\varepsilon_r = \varepsilon/\varepsilon_0$. A pair of parallel wires of diameter d and spacing a > 2.5 d has the impedance

$$Z = 120 \,\Omega \cdot \frac{1}{\sqrt{\varepsilon_r}} \ln \frac{2a}{d} \,. \tag{7.8}$$

This yields the impedance of a wire at a spacing h above a ground plane,

$$Z = 60 \ \Omega \cdot \frac{1}{\sqrt{\varepsilon_r}} \ln \frac{4h}{d} \,. \tag{7.9}$$

A common configuration utilizes a thin rectangular conductor above a ground plane with an intermediate dielectric (microstrip). Detailed calculations for this and other transmission line configurations are given by Gunston.*

7.3. Synchrotron radiation (CGS units)

For a particle of charge e, velocity $v = \beta c$, and energy $E = \gamma mc^2$, traveling in a circular orbit of radius R, the classical energy loss per revolution δE is

$$\delta E = \frac{4\pi}{3} \frac{e^2}{R} \beta^3 \gamma^4 .$$
 (7.10)

For high-energy electrons or positrons ($\beta\approx 1),$ this becomes

$$\delta E \text{ (in MeV)} \approx 0.0885 \ [E(\text{in GeV})]^4 / R(\text{in m})$$
. (7.11)

For $\gamma \gg 1$, the energy radiated per revolution into the photon energy interval $d(\hbar\omega)$ is

$$dI = \frac{8\pi}{9} \alpha \gamma F(\omega/\omega_c) d(\hbar\omega) , \qquad (7.12)$$

where $\alpha = e^2/\hbar c$ is the fine-structure constant and

$$\omega_c = \frac{3\gamma^3 c}{2R} \tag{7.13}$$

is the critical frequency. The normalized function F(y) is

$$F(y) = \frac{9}{8\pi}\sqrt{3} \ y \ \int_{y}^{\infty} \ K_{5/3}(x) \ dx \ , \tag{7.14}$$

where $K_{5/3}\left(x\right)$ is a modified Bessel function of the third kind. For electrons or positrons,

$$\hbar\omega_c (\text{in keV}) \approx 2.22 \ [E(\text{in GeV})]^3 / R(\text{in m}) .$$
 (7.15)

Fig. 7.1 shows F(y) over the important range of y.

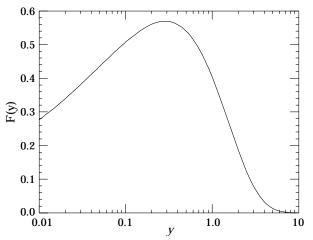


Figure 7.1: The normalized synchrotron radiation spectrum F(y).

For $\gamma \gg 1$ and $\omega \ll \omega_c$,

$$\frac{dI}{d(\hbar\omega)} \approx 3.3\alpha \, (\omega R/c)^{1/3} \quad , \tag{7.16}$$

whereas for

$$\frac{dI}{d(\hbar\omega)} \approx \sqrt{\frac{3\pi}{2}} \alpha \gamma \left(\frac{\omega}{\omega_c}\right)^{1/2} e^{-\omega/\omega_c} \left[1 + \frac{55}{72} \frac{\omega_c}{\omega} + \dots\right] \quad . \tag{7.17}$$

The radiation is confined to angles $\lesssim 1/\gamma$ relative to the instantaneous direction of motion. For $\gamma \gg 1$, where Eq. (7.12) applies, the mean number of photons emitted per revolution is

 $\gamma \gg 1$ and $\omega \gtrsim 3\omega_c$,

$$N_{\gamma} = \frac{5\pi}{\sqrt{3}} \alpha \gamma , \qquad (7.18)$$

and the mean energy per photon is

$$\langle \hbar \omega \rangle = \frac{8}{15\sqrt{3}} \hbar \omega_c \ . \tag{7.19}$$

When $\langle \hbar \omega \rangle \gtrsim O(E)$, quantum corrections are important.

^{*} M.A.R. Gunston. Microwave Transmission Line Data, Noble Publishing Corp., Atlanta (1997) ISBN 1-884932-57-6, TK6565.T73G85.

See J.D. Jackson, *Classical Electrodynamics*, 3^{rd} edition (John Wiley & Sons, New York, 1998) for more formulae and details. (Note that earlier editions had ω_c twice as large as Eq. (7.13).